

TITLE: STATE COVERAGE TOOL

FIELD OF INVENTION

5 The present invention relates to testing the state coverage of Hardware Design Language (HDL) simulations of integrated circuit designs.

BACKGROUND TO INVENTION

10 When verifying a digital hardware design specified using an HDL, the verification engineer is often interested to know whether the design has passed through a certain configuration of internal component signals, known as a "state".

15 Modern HDLs can be executed on simulation systems to produce traces of how the signals change over time. In the case of processor cores, for example, the simulation can effectively run the machine code accepted by the design. When verifying a module within the design, a test harness known as a testbench can be used to supply the correct stimuli to the design.

25 The signal traces can often be very long, and in a format that makes ascertaining what has actually happened relatively difficult for an engineer to ascertain, especially, where the state of interest constitutes the behaviour of multiple components over time.

30 SUMMARY OF INVENTION

According to the invention, there is provided a method of verifying a digital hardware design simulated in a hardware design language (HDL), the method including the steps of:

defining at least one state to be verified, the state including a signal value for each of a plurality of components within the hardware design;

applying a test to the hardware design;

5 generating traces of internal signals within the hardware design during the test, each trace including signal data and time data and including at least the internal signals associated with the components;

10 processing the traces to ascertain whether the plurality of components simultaneously had the signal values associated with the state, thereby to ascertain whether the state was achieved.

15 Preferably, more than one state is defined, each state having a signal value for each of a plurality components of the hardware design. The traces are processed to ascertain, for each state, whether the corresponding plurality of components simultaneously had the signal values associated with the state, thereby to ascertain whether each of the states was achieved. More 20 preferably, the processing step includes ascertaining whether a predetermined sequence of states was achieved.

25 Preferably also, the processing step includes ascertaining whether a given state in a sequence was achieved within a predetermined time period after an earlier state in the sequence.

30 In a preferred form, the traces are pre-processed prior to the processing step, such that, for at least each of the components defined within each of the states, a value for the respective signals associated with those components exists for each time for which the traces are to be processed.

It is preferred that one or more of the signal values are values of a field associated with the corresponding component.

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BRIEF DESCRIPTION OF DRAWINGS

Figure 1 is a flowchart showing coverage for a test applied to an HDL model of digital hardware, 10 incorporating state coverage according to the invention;

Figure 2 is a flowchart showing the steps involved in state matching according to the invention; and

15 Figure 3 is a flowchart showing the steps involved in sub-state matching.

DETAILED DESCRIPTION OF PREFERRED AND OTHER EMBODIMENTS

20 Referring to Figure 1, an HDL model is generated in a manner known to those skilled in the art. The HDL model includes a description of each of the various components in the model, such as registers, logical gates or more complex componentry. The model is used to predict the
25 behaviour of a model before it is committed to silicon.

In the flowchart shown in Figure 1, a first sub-process 102 is performed, in which a predetermined test is applied to the HDL model. As the test runs, signals associated with each of the components are recorded, with reference to a system clock that maintains system timing. Typically, this will involve monitoring of each component and recording changes that occur to signals associated with one or more of that component's aspects. For

example, in the case of writing to a register, data will be generated stating that the value of the register has changed, recording the new value and the time at which the change occurred.

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The result after the test has finished running is a dump file, which describes how every signal within the design changed over time. The dump file is in a Value Change Dump format, which includes ASCII representations of 10 signals, values and times.

POINTERS TO THE DESIGN

At this stage, it is desirable to run a pre-processing routine that removes data related to components that are not of interest. The pre-processing routine also groups signals together by name, and represents signals more than one bit wide as corresponding vectors of the correct size. This improves readability whilst limiting the size of the amount of data that needs to dealt with in subsequent steps.

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It is also desirable to pre-process the data in such a way that the change-based data is duplicated for each clock value. So, if a value of a register changes at clock cycle 10, and then again at clock cycle 50, and 25 there are 100 clock cycles in the test, the data is translated from two entries showing the new value and the time it changed into an entry for each of the 100 clock cycles along with the value of the register during that cycle. This leads in many cases to a drastic increase in 30 the amount of data stored, but it correspondingly simplifies the subsequent matching of states.

A list of coverage tests (CTs) is then selected in step 104. It will be appreciated that each CT can be as

simple as a single state that is to be tested for, and only a single test might be selected. In other cases, multiple tests with one or more sequences of states to be tested for might be selected.

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One of the CTs is then selected in step 106 to be tested for. Sub-process 108 is then run, which is described in more detail in relation to Figures 2 and 3. An assessment 110 is made as to whether the CT passed, and 10 in the event it did, this result is stored 112 in a log for later review. A further assessment 114 is then made as to whether any further CTs remain to be tested for. If so, the procedure returns to step 106 and continues. If not, then CTs that passed are removed 116 from the CT 15 list, and an assessment 118 made as to whether any CTs remain. If so, then the test as a whole is failed 120, and if not, then it passes 122.

Turning to Figure 2, sub-process 108 is shown in more 20 detail. Here it is assumed that there are multiple states to be tested for in each CT. To begin with, a state is selected 202 for testing against the pre-processed data, which includes an entry for every clock cycle for each component of interest, the entry including 25 the value of signal of the component at that time. State matching then takes place in step 204.

Step 204 is a looped set of sub-steps. The first step is to evaluate 206 all of the signals of the components of 30 interest at a particular time. In the event that all of the signals match 208 those defined in relation to the state being considered, then the state is found 210. If the state is not found at that time, then an assessment 212 is made as to whether there is still time for that

state to occur. If such time is not available, the state is deemed not found 214. If such time is available, then the time is advanced 216 (typically by a single clock period) and the signals at the new time again evaluated 5 208 to ascertain whether the state has occurred. This loop continues until no further time is available or it is determined that the state has been reach.

On the issue of available time, it will be appreciated 10 that the HDL model of digital hardware does not behave in an identical way to the actual hardware itself. In some cases, it is the movement of data and control signals via a control port that must be waited for. Also, latency can be defined by the programmed to allow time for data 15 to ripple through multiple components. It is intrinsically difficult to accurately model such cascading effects, and so it may be necessary to introduce windows of allowable latency to prevent tests that perform correctly being rejected because states do 20 not happen at precisely the anticipate time

Once the state-matching step 204 is finished, an assessment 218 is made as to whether the state was found. If not, then the CT failed 220 and the procedure 25 continues as per Figure 1. If the state was found, then sub-state matching takes place 222, as described in more detail in relation to Figure 3.

If sub-states are not found 224, then the CT failed 220 30 and the procedure continues as per Figure 1. If the required sub-states are found, however, the next basic state can be selected 226 and the procedure returns to the state matching routine 204. This continues until all

states and sub-states have been considered, and then the procedure continues as per Figure 1.

Sub-state matching is used to reduce the likelihood of rejection of an acceptable test due to latency issues. The difference between a *state* and a *sub-state* is that a state is the set of signals that should be observed as all having the predetermined values at one given time. These signals will typically be those that are updated by the device itself, rather than as a response to a deliberate write to the relevant components due to implementation of the test code. Sub-states are generated in response to statements in the test that alter some internal state of a component, which introduces latency. Part of the declaration of a sub-state (discussed in more detail below) is a DELTA line that specifies the maximum allowable latency before the sub-state fails, causing the test to fail. Generally, sub-states are required when dealing with writes to control registers and similar operations.

The preferred embodiment requires that a state must first be ascertained as existing, in that the required signal not affected by the test code should all have the correct values at one particular time, t . Having matched the basic state, the procedure then start searching for sub-states, from time t . Each sub-state is searched for from this time. If all the sub-states are found within the given delta, then the whole state has passed. If they are not all found, the then test fails.

An example of where latency could be problematic will now be described. The example assumes that the coverage task is generated from an abstract test, and that the concrete

test is running in an environment where there is only one target port into the peripheral. This means that only one address or peripheral register can be accessed at a time.

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The peripheral module has a set of control registers COUNT and ENABLE. COUNT is initially programmed by the test and automatically updated once the test has started. ENABLE is non-volatile, and retains its value until is updated by the test.

A test might specify the initialisation of all the control registers in the device into one state at the beginning of the test. The corresponding concrete test will need to sequentially set the control registers over several clock cycles, since the peripheral has only one target port. Provided the registers or peripheral state are not volatile, the state will be complete when the last transfer is made.

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On subsequent states it is not desirable to try reinitialising the control registers that are automatically modified, because this would mean that all that is being tested is the ability of the control register to accept reads and writes, and latency would probably affect the outcome of the test. However, it is likely that the subsequent states are going to try to program other control registers in the device as well as observing the value of the auto-modified register.

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There will always be a latency associated with writing to a control register, which means that the state actually spreads over multiple clock cycles. For example, assume that the COUNT register contains the value 10 when the

test attempts to zero the enable register to suspend the device. Due to latency, it is possible that the COUNT register will be decremented to 9 before the disable actually takes place.

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If the state were to be interpreted as being a time at which COUNT is equal to 10 and the device disabled, the test would be counted a failure.

10 Turning to Figure 3, the steps for sub-state matching 222 are shown. First a sub-state is chosen 302 from a list of sub-states to be searched for. A counter, d , is initialised 304 to zero, and the signals(s) of interest are evaluated 306 at time $t + d$. If the signals match 308 the state that is being searched for, then the sub-state is found 310. If the sub-state is not found, it is determined 312 whether there is any time remaining. If no time remains, then the sub-state is not found 314. If there is time remaining, d is incremented 316. If the subsequent value of d exceeds a predetermined maximum, then the state is deemed not found 314. When the maximum for d is not reached at this stage, then the subroutine goes back to evaluating 306 signals at the new time.

25 Once the sub-state has been found or the time has run out, an assessment 320 is made, and if the sub-state was not found, the routine is deemed to fail 322. If the sub-state was found, the time is reset 324 to t , and it is determined 326 whether there are any more sub-states 30 to search for. If not, then the sub-state search routine is completed 328. However, if there are more sub-states to be searched for, then the next one is selected 330 and the routine returns to step 304. The loop is repeated until all sub-states have been searched for, and then

continues with the procedure outlined in relation to Figure 1.

5 Signals can be defined in each state in terms of the signal per se, or with reference to a field or mask. The following is an example signal definition in an HDL

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10 SIGNAL : testbench.bclk  
15 SIGNAL  
    :testbench.dmac_dmacontroller_qbinst.u_dmaccr.cr_com_sw_v  
    iew  
15 SIGNAL  
    :testbench.dmac_dmacontroller_qbinst.u_dmaccr.cr_ch0ctl_s  
    w_view  
20 FIELD : master_enable,  
    testbench.dmac_dmacontroller_qbinst.u_dmaccr.cr_com_sw_v  
    iew, .*([01xz])[01xz]{3}  
25 FIELD : priority,  
    testbench.dmac_dmacontroller_qbinst.u_dmaccr.cr_com_sw_v  
    iew, .*([01xz])  
30 FIELD : transfer_enable0,  
    testbench.dmac_dmacontroller_qbinst.u_dmaccr.cr_ch0ctl_sw  
    _view, .*([01xz])[01xz]{12}  
30 FIELD : resource_select0,  
    testbench.dmac_dmacontroller_qbinst.u_dmaccr.cr_ch0ctl_sw  
    _view, .*([01xz){4})[01xz]{7}
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5 It will be seen that several fields can be extracted from one signal, and any state that is being searched for can be defined in terms of signals and/or their parts (e.g., fields). This can also be achieved by masking, where the entire signal data is looked at, but a mask applied to ensure that only bits that are of interest are compared with the desired state.

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The method of the invention drastically simplifies the task of hardware design engineers by allowing easy and relative efficient assessment of state coverage.

15 Although the invention has been designed with reference to specific examples, it will be appreciated that the invention can be embodied in many other forms.

Having thus described at least one illustrative embodiment of the invention, various alterations, modifications, and improvements will readily occur to those skilled in the art. Such alterations, modifications, and improvements are intended 20 to be within the spirit and scope of the invention. Accordingly, the foregoing description is by way of example only and is not intended as limiting. The invention is limited only as defined in the following claims and the equivalents thereto.

25 What is claimed is: